Notes ! (1) 132 = 5.65 2 IGNORE X'S & O'S INP. 6. ANSWER

ECGR4124 Digital Signal Processing

Exam 2

Spring 2017

Name:	\circ

LAST 4 NUMBERS of Student Number: _____

Do NOT begin until told to do so
Make sure that you have all pages before starting
NO TEXTBOOK, NO CALCULATOR, NO CELL PHONES/WIRELESS DEVICES
Open handouts, 2 sheet front/back notes, NO problem handouts, NO exams, NO quizzes
DO ALL WORK IN THE SPACE GIVEN
Do NOT use the back of the pages, do NOT turn in extra sheets of work/paper
Multiple-choice answers should be within 5% of correct value
Show ALL work, even for multiple choice

ACADEMIC INTEGRITY:

Students have the responsibility to know and observe the requirements of The UNCC Code of Student Academic Integrity. This code forbids cheating, fabrication or falsification of information, multiple submission of academic work, plagiarism, abuse of academic materials, and complicity in academic dishonesty.

Unless otherwise noted:

 $e^{j\theta} = \cos(\theta) + j \sin(\theta)$

F{} denotes Discrete time Fourier transform {DTFT, DFT, or Continuous, as implied in problem} F{} denotes inverse Fourier transform

 ω denotes frequency in rad/sample, Ω denotes frequency in rad/second

* denotes linear convolution, (N) denotes circular convolution

 $x^*(t)$ denotes the conjugate of x(t)

Useful constan	Control of the second of the s	
e ≈ 2.72		≈ 3.14
$e^2 \approx 7.39$		≈ 54.6
$e^{-0.5} \approx 0.607$		$^{25} \approx 0.779$
1/e ≈ 0.37		≈ 1.41
e ⁻² ≈ 0.135		≈ 1.73
e ⁻⁴ ≈ 0.0183 √		≈ 2.22
√7 ≈ 2.64		10 ≈ 3.16
$ln(2) \approx 0.69$		$n(4) \approx 1.38$
$log_{10}(2) \approx 0.3($		$\log_{10}(3) \approx 0.48$
$log_{10}(10) \approx 1.0$		$\log_{10}(0.1) \approx -1$
$1/\pi \approx 0.318$		sin(0.1) ≈ 0.1
tan(1/9) ≈ 1/9		$\cos(\pi/4) \approx 0.71$
cos(A)cos(B)	= 0.5 cos(A - B) + 0.5 cos(A +	· B)

- 1. A continuous-time sinusoid of 250 Hz sampled at 1000 samples/second has a corresponding discrete-time frequency of ω =
 - a) $\pi/4$ rad/sample
- b) $\pi/2$ rad/sample c) π rad/sample
- d) none above

- 2. The impulse response of a system is $h[n] = \delta[n] \delta[n-1]$, the frequency response is $H(\omega)$ =
 - a) $sinc(2\omega) e^{-j\omega}$
- b) $e^{-j\omega} e^{-j2n\omega}$ c) 1 $e^{-j\omega}$ d) none above

- 3. Pre-warped bilinear transform filter designs suffer from aliasing.
 - a) True

- b) False
- Circle the causal BIBO stable impulse response below.
- a) $h[n] = e^{-j2n\pi} u[n]$ b) h[n] = j u[n] c) $h[n] = e^{-2n\pi} u[n]$
- d) none above

- 5. If a filter has z-transform $H(z) = \frac{z^3 + 2z + 2}{5z^3}$; |z| > 0, then the filter is
 - a) FIR
- b) IIR
- c) two-sided
- d) none above

Y(z)=15 X(z)=+= Y(z)z

- 6. The dc response of a system with difference equation y[n] = 15 x[n-1] + 0.5 y[n-2] is:
 - a) 7.5
- b) 10
- c) 30
- d) none above

- 7. If a filter has $H(z) = \frac{2z-2}{5z+3}$ and ROC |z| > 0.6 then, the response at $\omega = \pi$ is $H(\omega)|_{\omega = \pi} = H(\pi) = 0.6$
- c) 1/2
- (d)2)
- e) none above

 $W=T=2=-1=)\frac{-2-2}{-5+2}=\frac{-4}{-7}=2$

- 8. If $Y(z)=1+z^{-1}$; |z|>0, and $X(z)=1+z^{-1}$; |z|>0, then the convolution x[n]*y[n]=
- d) none above

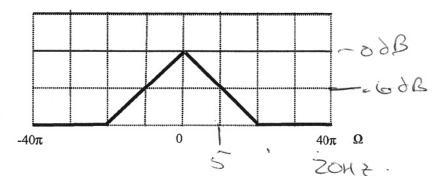
y (n)={1,1,...} *=> 011...

- 9. If a filter is to be designed using the windowing methods where h[n]=w[n] h_d[n], and where $h_d[n]=2^{-n}$ u[n], which of the following windows has the lowest relative peak sidelobe:

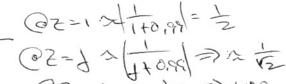
- 10. The magnitude of the frequency spectrum of a signal $|X(\Omega)|$ is given below. The 6 dB bandwidth of the signal is
 - a) 5 Hz)
- b) 10 Hz
- c) 15 Hz
- d) 20 Hz
- e) none above

 $|X(\Omega)|$ 1

0



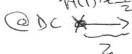


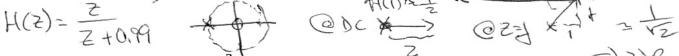


11. For a system with h[n] = (-0.99)ⁿ u[n], the 3 dB bandwidth is (2 > -1 > 1)

- a) $\pi/8$ rad/sample b) $\pi/4$ rad/sample







12. For a system with $H(z) = \frac{z^2+1}{z^2}$; |z|>0, the corresponding 4-point DFT is X[k]=

- a) {0, 2, 0, 2}
- b) {0, -2, 0, -2 } (c) {2, 0, 2, 0}
- d) none above



$$= \frac{1}{2} \left\{ \frac{1}{1} \right\} \frac{1}{2} \frac{1}{1} \left(\frac{1}{1} \right)^{2}$$

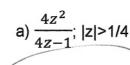
$$= \left\{ \frac{(1^{2}+1)}{1} \frac{J^{2}+1}{J^{2}} \frac{(1)^{2}+1}{(1)^{2}} \frac{(-1)^{2}+1}{(-1)^{2}} \right\} = \left\{ 2,0,2,0 \right\}$$

13. The stable 2-sided LTI system with z-transform $H(z) = \frac{2z+0.5}{(z+2)(3z+1)}$ has an ROC of

- a) 1/3 < |z| < 2
- b) 1/3 < |z| < 1/2
- c) 1/2 < |z| < 3
- d) none above



14. The z-transform of h[n]= $4^{-(n-1)}$ u[n-1] is H(z) =



b)
$$\frac{4z^{-1}}{4z-1}$$
; $|z| > 1/4$

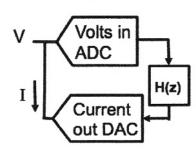
$$(c)\frac{4}{4z-1}; |z|>1/4$$

d) none above

(4) => = Z-1/4 => = 4 = 42-1

15. In the system below, $H(z) = -5.0 (1-z^{-1})$ and the clock period is T=1 nanosecond, impedance of the circuit looks like the impedance a capacitor of

- a) -5 nF
- b) 5 nF
- c) -5 uF
- d) none above



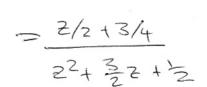
i=can/1= ~ = (Non)-Non)

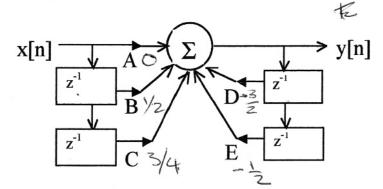
0--5T= -5X10

4/10

5 Points Each, Circle the Best Answer

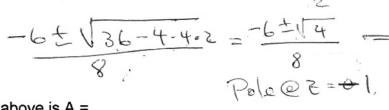
Let the following causal system be initially at rest, and let h[n] be the impulse response, with z-transform $H(z) = \frac{2z+3}{4z^2+6z+2}$.





16. The system is BIBO stable

- a) True
- b) False



17. The filter coefficient A in the block diagram above is A =

- a) -1/2
- b) 0
- c) 1/2
- d) 2/3
- e) none above

18. The filter coefficient C in the block diagram above is C =

- a) -1/2
- b) 1/2
- c) 3/4
- d) 3
- e) none above

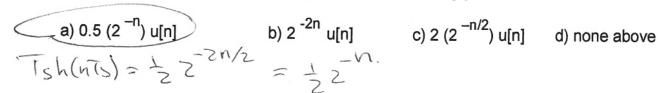
19. The filter coefficient E in the block diagram above is E =

- a) -1/2
- b) -1/3
- c) 1/2
- d) 2/3
- e) none above

20. The second point of the impulse response is, h[1] =

- a) 0
- b) 1/3
- c) 1/2
- d) 3/2
- e) none above

- 5 Points Each, Circle the Best Answer
- 21. A 2 sample/second filter with impulse response h[n] is constructed using the impulse invariance method for $h(t) = 2^{-2t} u(t)$. Given this filter, h[n]=



22. The difference equation of a causal system is 3y[n] = 5x[n] + y[n-1]. The z-transform of the system is H(z)=

A)
$$\frac{5z}{3z-1};|z|>1/3$$

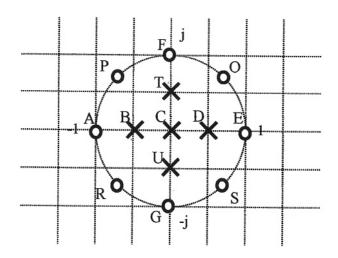
c)
$$\frac{5z}{z-3}$$
; $|z| > 3$

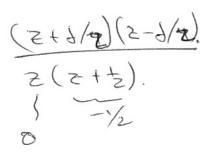
b)
$$\frac{5z}{3z+1}$$
; $|z| > 1/3$

d)
$$\frac{5z}{z+3}$$
; $|z| > 3$

b)
$$\frac{5z}{3z+1}$$
; $|z|>1/3$ $(3-2^{-1}) = 5 \times (2)$
 $(3-2^{-1}) = 5 \times (2)$
d) $(3-2^{-1}) = 5 \times (2)$
 $(3-2^{-1}) = 3 \times (2)$
e) none above

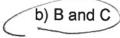
The following questions are for an causal LTI filter with $H(z) = \frac{z^2 + 1/4}{z^2 + z/2}$.



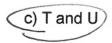


Ignore the X's and O's in the plot. Only be concerned with their locations. they do not necessarily indicate a pole or a zero at each location. only their coordinates are important in the questions below.

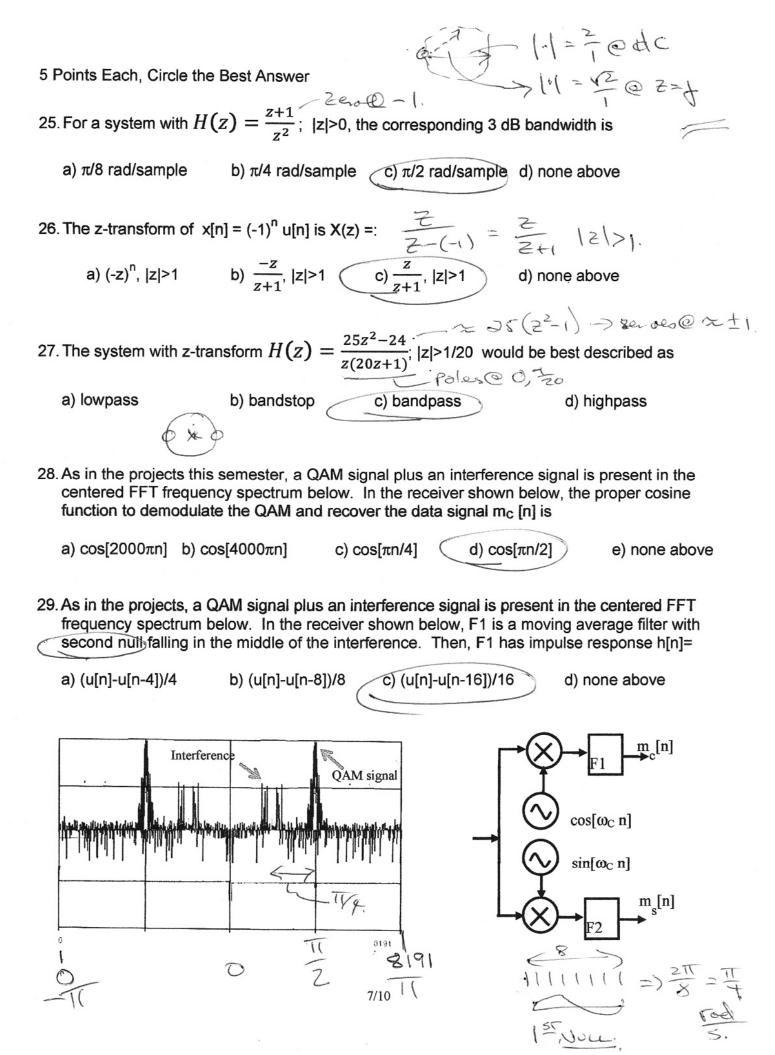
- 23. The proper locations of the poles of H(z) are at locations (sorry for any confusion)
 - a) C and D



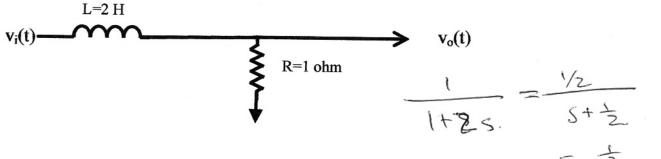
- c) T and U
- d) none above
- 24. The proper locations of the zeroes of H(z) are at locations
 - a) T and C
- b) B and C



d) none above



In the following questions, a discrete-time filter is to be designed using the impulse invariance method. The sample rate of the digital system is 2 samples second. The discrete-time filter is to replace the causal continuous-time filter below, with H(s) = Vo(s)/Vi(s).



30. For the above continuous-time circuit, H(s) = Vo(s)/Vi(s) =

a)
$$\frac{2s}{1+s}$$

b)
$$\frac{1}{1+s}$$

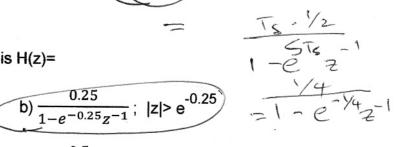
b)
$$\frac{1}{1+s}$$
 c) $\frac{2}{s+1}$

d) none above

31. The z-transform of the discrete-time filter is H(z)=

a)
$$\frac{1}{1-e^{-1}z^{-1}}$$
; $|z| > e^{-1}$

b)
$$\frac{0.25}{1 - e^{-0.25}z^{-1}}$$
; $|z| > e^{-0.25}$



c)
$$\frac{0.2}{1-e^{-0.2}z^{-1}}$$
; |z|> $e^{-0.2}$

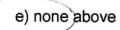
d)
$$\frac{0.5}{1-e^{-0.5}z^{-1}}$$
; |z|> $e^{-0.5}$

e) none above

32. For the discrete-time filter, the first point in the impulse response of the filter is h[0]=

a) 1/2

- b) 1

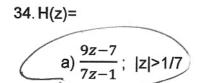


33. If the filter design is modified for a new sampling rate of 2000 samples/second, the resulting discrete-time filter is BIBO stable.

- a) True
- b) False

A discrete-time filter is to be designed using the bilinear transform method. The sampling rate of the digital system is 2 samples/second. The continuous-time filter is given as:

$$H(s) = \frac{2s+1}{s+3} \implies \frac{2^{\frac{2}{7}s^{\frac{2-1}{5}}+1}}{\frac{2}{7}s^{\frac{2-1}{5}}+3}$$



c)
$$\frac{5z-1}{6z+2}$$
; |z|>1/3

e) none above

b)
$$\frac{5z-3}{5z+1}$$
; $|z|>1/5$ $\frac{8}{2+1}+1$ $\frac{1}{2+1}+3$

d)
$$\frac{3z+1}{3z+2}$$
; $|z|>2/3$ = $8z-8+2+1$
 $4z-4,432+3$

$$= \frac{42-4.132.13}{42-4.132.13}$$

$$= \frac{.92-7.}{72-1.}$$

35. At frequency $\omega=0$ rad/sample, the frequency response of the discrete-time filter $|H(\omega)|$ most nearly equals

a) 4/9

- b) 1/2 (c) 1/3 d) 4/5
- e) none above

$$\frac{25f1}{5f3}$$
 $CS=0=\frac{1}{3}$

36. At frequency $\omega = \pi$ rad/sample, the frequency response of the discrete-time filter $|H(\omega)|$ most nearly equals

- a) 0
- b) 1/2
- c) 3/2
- e) none above

$$\frac{\partial(\infty)+1}{\infty+3}=2$$

37. If the filter is changed to a prewarped bilinear design with a prewarp frequency of 1 Hz, the dc frequency response of the discrete-time filter $|H(\omega)|_{\omega=0}$ would then equal

a) arctan(1)



- c) $7/\arctan(2\pi)$
- d) 7/6
- e) none above



The following questions refer to the Java class below and the program main(). JAVA NOSED

```
public class Green {
  private int a; private int b;
```

```
public Green(int aa, int bb)
      { a=aa; b=bb; }
```

public void equals(Green c)

 $\{ \text{this.a} = \text{c.a}; \text{this.b} = \text{c.b}; \}$

public void fn(Green c)

```
{ this.a \neq c.a + this.a;
```

this.b \Rightarrow this.a - c.a; }

public void gg()

 $\{ \text{this.a} = \text{this.a} + 2;$

this.b = this.a - 1;}

public static void main(String[] args) {

Green x = new Green (1,2);

Green y = new Green (2,1);

Green z = new Green (1,1);

int xx=1,yy=2,zz \pm 3;

x.fn(y);

z.gg();

38. At the end of the main program, x\a

- a) 1
- b) 2

d) None above

39. At the end of the main program, x.b=

- b) 2
- c) 3
- d) None above

40. At the end of the main program, z.a=

- a) 1
- c) 3
- d) None above