

5 Points Each, Circle the best answer

1. Assuming a 200 ohm voltage source, the reflection coefficient of a 100 ohm load is $\Gamma =$
 a) -1/2 **b) -1/3** c) 1/4 d) 1/3 e) none above

$$\Gamma = \frac{Z_L - Z_S}{Z_L + Z_S} = \frac{100 - 200}{100 + 200} = \frac{-1}{3}$$

2. In a 30 ohm system, the return loss of a 50 ohm load is
 a) -3 dB b) 3 dB c) 6 dB **d) 12 dB** e) none above

$$-10 \log_{10}(|\Gamma|^2) = -10 \log_{10} \left(\left| \frac{50 - 30}{50 + 30} \right|^2 \right) = -10 \log_{10} \left(\left| \frac{20}{80} \right|^2 \right) = -10 \log_{10} \left(\frac{1}{16} \right) = 12 \text{ dB}$$

3. If propagation constant $\gamma = 2 + j\pi/2$ in a transmission line, the wavelength $\lambda =$
 $= \alpha + j\beta$
 a) $1/\pi$ m b) 2 m **c) 4 m** d) 8 m e) none above

$$\lambda = \frac{2\pi}{\beta} = \frac{2\pi}{\pi/2} = 4$$

4. If propagation constant $\gamma = j2\pi$ in a transmission line, the line is lossless.
 $= \alpha + j\beta \Rightarrow \alpha = 0$
a) True b) False

5. In a right-handed LC transmission line with $L_R = 300$ nH/m and $C_R = 2000$ pF/m, the wave equation would be

a) $\frac{\partial^2 v(x,t)}{\partial x^2} = 150 \frac{\partial^2 v(x,t)}{\partial t^2}$

b) $\frac{\partial^2 v(x,t)}{\partial x^2} = \frac{6}{10^{16}} \frac{\partial^2 v(x,t)}{\partial t^2}$

c) $\frac{\partial^2 v(x,t)}{\partial x^2} = \frac{10^{16}}{6} \frac{\partial^2 v(x,t)}{\partial t^2}$

d) none above

$$\frac{\partial^2 V}{\partial x^2} = L_R C_R \frac{\partial^2 V}{\partial t^2} \Rightarrow \frac{300}{10^9} \cdot \frac{2000}{10^{12}} = \frac{6}{10^{16}}$$

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6. A coaxial line with 5.4 mm outer radius and 2 mm inner radius with $\epsilon_r=16$, $\mu_r=1$ has an impedance in ohms of $Z_0=$

- a) 15 b) 30 c) 60 d) 90 e) none above

$$Z_0 = \sqrt{\frac{\mu}{\epsilon}} \frac{\ln(b/a)}{2\pi} = \sqrt{\frac{\mu_0}{16\epsilon_0}} \frac{\ln(5.4/2)}{2\pi} = 94 \cdot \frac{1}{2\pi} = 15$$

7. In a transmission line with propagation constant $\gamma = 2 + j\omega^3/6$, the group velocity is $v_g=$

- a) $2/\omega^2$ m/s b) $4/\omega^2$ m/s c) $2\omega^2$ m/s d) $j4\omega^2/3$ m/s e) none above

$$v_g = \frac{d\omega}{d\beta} = \frac{1}{d\beta/d\omega} = \frac{1}{d(\omega^3/6)/d\omega} = \frac{1}{3\omega^2/6} = 2/\omega^2$$

8. For a system with 50-ohm S-parameters $S = \begin{bmatrix} 0.5 & -0.25 \\ -1.5 & 0.5 \end{bmatrix}$ and a 150 ohm load, the input reflection coefficient is $\Gamma_{in} =$

- a) 0.2 b) 0.33 c) 0.75 d) 0.9 e) none above

$$\Gamma_{in} = S_{11} + \frac{S_{21}S_{12}\Gamma_L}{1+S_{22}\Gamma_L} = \frac{1}{2} + \frac{(-\frac{3}{2})(-\frac{1}{4})(\frac{1}{2})}{1 - (\frac{1}{2})(\frac{1}{2})} = \frac{1}{2} + \frac{3/16}{3/4} = \frac{3}{4}$$

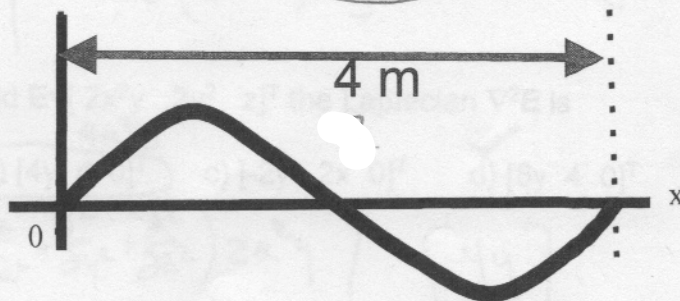
9. For a system with 50-ohm S-parameters $S = \begin{bmatrix} 0.5 & 0 \\ 8 & 0 \end{bmatrix}$ with a 50-ohm source and a 150 ohm load, the magnitude of the transducer gain is $|g_T|=$

- a) 12 b) 24 c) 48 d) 96 e) none above

$$g_T = \frac{|S_{21}|^2 (1-|\Gamma_S|^2)(1-|\Gamma_L|^2)}{|(1-S_{11}\Gamma_S)(1-S_{22}\Gamma_L) - S_{21}S_{12}\Gamma_S\Gamma_L|^2} = \frac{8^2(1-0)(1-\frac{1}{4})}{|(1-\frac{1}{2}\cdot 0)(1-0\cdot\frac{1}{2}) - 8\cdot 0\cdot 0\cdot\frac{1}{2}|^2} = \frac{64\cdot 3}{4} = 48$$

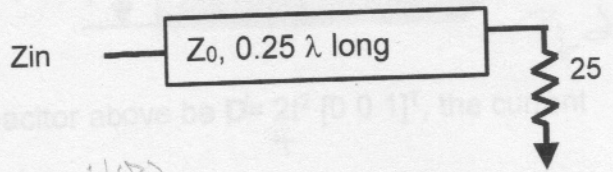
10. The following sinusoidal waveform is observed on a transmission line at time $t=0$, where the transmission line phase velocity is known to be $v_p=10^8$ m/s. The frequency of the sine wave is $f=$

- a) $20/\pi$ MHz b) $10\pi/4$ MHz c) 25 MHz d) 50 MHz e) none above



$$v_p = \lambda f \Rightarrow f = v_p / \lambda = \frac{10^8}{4}$$

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11. For the circuit above to have $Z_{in}=100$ ohms, the line impedance in ohms must be $Z_0=$

- a) 20 b) 40 **c) 50** d) 75 e) none above

$$Z_0 = \sqrt{\frac{100}{25} \cdot 25^2} = \sqrt{2500} = 50$$

12. For the two electric fields $\mathbf{E}_1=[2 \ 2 \ 2]^T$ and $\mathbf{E}_2=[1 \ 2 \ 3]^T$ the cross product $\mathbf{E}_1 \times \mathbf{E}_2$ is

- a) $[2 \ -4 \ 6]^T$ b) $[4 \ -2 \ 2]^T$ **c) $[2 \ -4 \ 2]^T$** d) $[-2 \ 4 \ -2]^T$ e) none above

$$\begin{vmatrix} \hat{x} & \hat{y} & \hat{z} \\ 2 & 2 & 2 \\ 1 & 2 & 3 \end{vmatrix} = \hat{x}(2 \cdot 3 - 2 \cdot 2) - \hat{y}(2 \cdot 3 - 2 \cdot 1) + \hat{z}(2 \cdot 2 - 2 \cdot 1)$$

$$= 2\hat{x} - 4\hat{y} + 2\hat{z}$$

13. For the voltage potential scalar field $V=x^3z$ the gradient ∇V is

- a) $[3x^2z \ 0 \ x^3]^T$** b) $[3x^2y \ x^3 \ 0]^T$ c) $[x^3z \ x^2 \ y]^T$ d) $[z \ 0 \ x^3z]^T$ e) none above

$$\begin{bmatrix} \partial/\partial x \\ \partial/\partial y \\ \partial/\partial z \end{bmatrix} x^3z = \begin{bmatrix} 3x^2z \\ 0 \\ x^3 \end{bmatrix}$$

14. For the electric field $\mathbf{E}=[2x^2y^2 \ 0 \ 0]^T$ the curl $\nabla \times \mathbf{E}$ is

- a) $[4xy \ 0 \ 0]^T$ **b) $[0 \ 0 \ -4x^2y]^T$** c) $[0 \ 0 \ -8x^2y]^T$ d) $[8xy \ 0 \ 0]^T$ e) none above

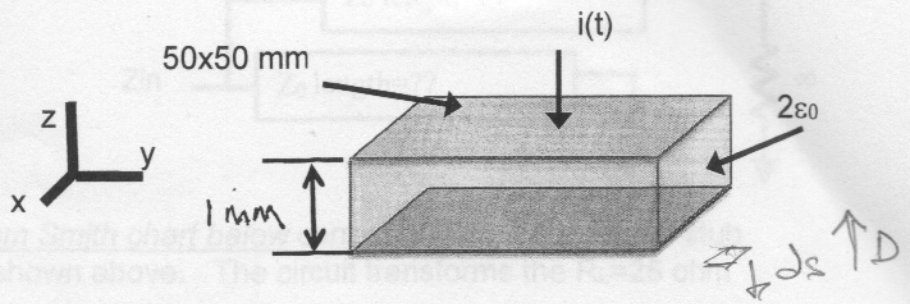
$$\begin{vmatrix} \hat{x} & \hat{y} & \hat{z} \\ \partial/\partial x & \partial/\partial y & \partial/\partial z \\ 2x^2y^2 & 0 & 0 \end{vmatrix} = \hat{x}(0-0) - \hat{y}(0-0) + \hat{z}(0-4x^2y)$$

15. For the electric field $\mathbf{E}=[2x^2y \ 2y^2 \ z]^T$ the Laplacian $\nabla^2 \mathbf{E}$ is

- a) $[2xy \ x \ 0]^T$ **b) $[4y \ 4 \ 0]^T$** c) $[-2y \ 2x \ 0]^T$ d) $[8y \ 4 \ 0]^T$ e) none above

$$\begin{bmatrix} \nabla^2 (2x^2y) \\ \nabla^2 (2y^2) \\ \nabla^2 z \end{bmatrix} = \begin{bmatrix} (\frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + \frac{\partial^2}{\partial z^2}) 2x^2y \\ \frac{\partial^2}{\partial x^2} 2y^2 \\ \frac{\partial^2}{\partial x^2} z \end{bmatrix} = \begin{bmatrix} 4y \\ 4 \\ 0 \end{bmatrix}$$

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16. Let the D-field between the two plates of the capacitor above be $\mathbf{D} = 2t^2 [0 \ 0 \ 1]^T$, the current in mA as shown into the top plate is $i(t) =$

- a) -10 t b) -20 t c) 10 t d) 20 t e) none above

$$I = \int \frac{dD}{dx} \cdot ds = -\frac{d}{dt} (2t^2) \cdot \left(\frac{50}{1000}\right)^2 = -4t \left(\frac{2500}{1000^2}\right) = \frac{-10t}{1000}$$

17. Let a 2-turn rectangular loop of 3 m^2 area in the x-y plane have a B-field along the z-axis of $\mathbf{B} = 2t^2 [0 \ 0 \ 1]^T$, the magnitude of the emf in volts is $|\text{emf}| =$

- a) 12 t b) 24 t c) 36 t d) 48 t e) none above

$$|\text{emf}| = \left| -\frac{N d\phi}{dt} \right| = \frac{2d}{dt} (2t^2 \cdot 3) = 24t$$

18. The capacitance of two $5 \times 5 \text{ mm}$ square metal plates separated by a 1 mm thick dielectric with $\epsilon = 9 \epsilon_0$ is

- a) 1 pF b) 1.5 pF c) 2 pF d) 2.5 pF e) none above

$$C = \frac{\epsilon A}{d} = \frac{9\epsilon_0 \left(\frac{5}{1000}\right)^2}{1/1000} = \frac{225}{1000} \cdot 8.85 \text{ pF/m} = 1.991 \text{ pF}$$

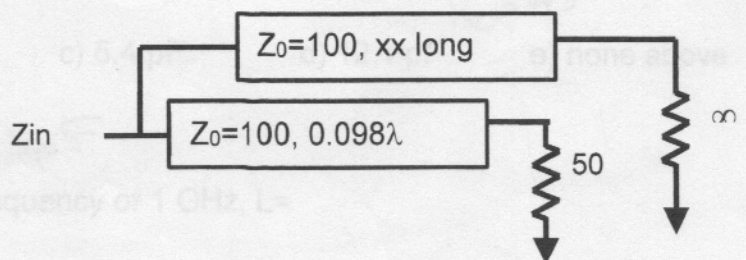
19. The inductance of a 5-turn inductor in vacuum with radius 2 mm and length 10 mm is

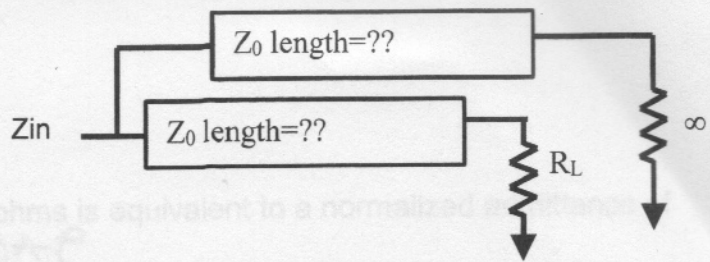
- a) 20 nH b) 40 nH c) 80 nH d) 160 nH e) none above

$$L = \frac{N^2 \pi r^2 \mu_r \mu_0}{d} = \frac{25 \pi \left(\frac{2}{1000}\right)^2 1257 \text{ nH/m}}{10/1000} = 39.5 \text{ nH}$$

20. In the single-stub match below, for $Z_{in} = 100 \text{ ohms}$, the length of the open-circuit stub is $xx =$

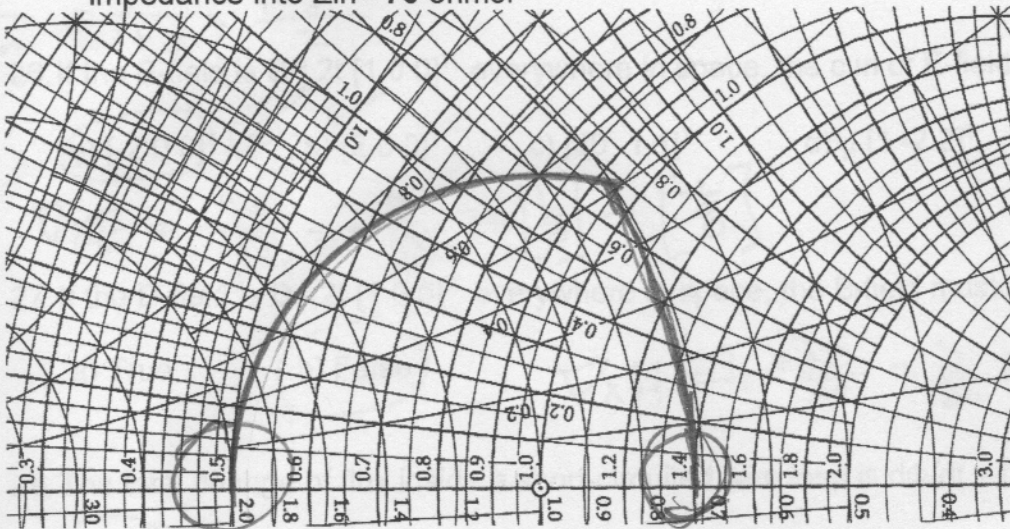
- a) 0.055λ b) 0.096λ c) 0.125λ d) 0.22λ



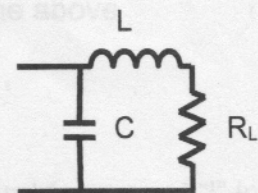
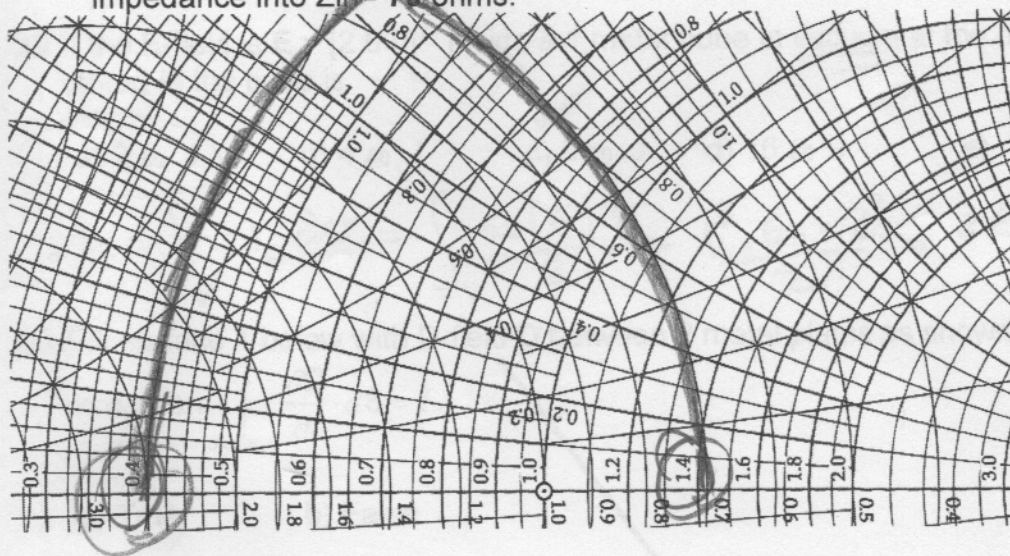


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21. Draw the paths on the $Z_0=50$ ohm Smith chart below corresponding to the single-stub matching network of the circuit shown above. The circuit transforms the $R_L=25$ ohm impedance into $Z_{in}=70$ ohms.



22. Draw the paths on the $Z_0=50$ ohm Smith chart below (for L and C) corresponding to the matching network of the circuit shown below. The circuit transforms the $R_L=20$ ohm impedance into $Z_{in}=70$ ohms.



$\frac{X_L}{50} \approx 0.65$
 $X_L = 32.5$
 $V_A = 1.1$
 $X_C = 0.91$
 $X_C = 45$

23. For the above impedance match at a frequency of 1 GHz, C=

- a) 1.2 pF **b) 3.6 pF** c) 5.4 pF d) 12.1 pF e) none above

$C = \frac{1}{\omega \cdot X_C} = \frac{1}{2\pi \cdot 10^9 \cdot 45} = 3.5 \text{ pF}$

24. For the above impedance match at a frequency of 1 GHz, L=

- a) 3.9 nH **b) 5.0 nH** c) 7.6 nH d) 14.3 nH e) none above

$L = \frac{X_L}{\omega} = \frac{32.5}{2\pi \cdot 10^9} = 5.2 \text{ nH}$

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25. An normalized impedance $Z_n = 1.0 + j 1.0$ ohms is equivalent to a normalized admittance of $Y_n =$

- a) $2 - j2$ b) $0.5 + j 0.5$ c) $0.5 - j 0.5$ d) $-2 - j$ e) none above

← SWAP

$$Y_n = \frac{1}{Z_n} = \frac{1}{1 + j} = \frac{1 - j}{2}$$

26. If the B-field is $\mathbf{B} = 2t [1 \ 0 \ 0]^T$ everywhere in space, the curl of E-field is $\nabla \times \mathbf{E} =$

- a) $[-2 \ 0 \ 0]^T$ b) $[2 \ 0 \ 0]^T$ c) $\mu [0 \ -1 \ 1]^T$ d) $\mu [1 \ -2 \ 2]^T$ e) none above

$$\nabla \times \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t} = -\frac{\partial}{\partial t} 2t \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix} = \begin{bmatrix} -2 \\ 0 \\ 0 \end{bmatrix}$$

27. If the D-field is $\mathbf{D} = 2t [1 \ 0 \ 0]^T$ everywhere in space, the B-field must be zero.

- a) True b) False

$$\nabla \times \mathbf{H} = \mathbf{J}_e + \frac{\partial \mathbf{D}}{\partial t} = \frac{\partial}{\partial t} 2t \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix} = \begin{bmatrix} 2 \\ 0 \\ 0 \end{bmatrix}$$

28. The total change of flux inside a short-circuited wire loop is $d\Phi/dt = 0$.

- a) True b) False

29. If the E-field is $\mathbf{E} = [2 \ 0 \ 0]^T$ inside a 1 meter cube in vacuum at the origin, the total energy stored in the cube is

- a) $2 \epsilon_0 \text{ J}$ b) $4 \epsilon_0 \text{ J}$ c) $8 \epsilon_0 \text{ J}$ d) $16 \epsilon_0 \text{ J}$ e) none above

$$W_E = \int_V w_E dV = \int_V \frac{\epsilon_0 |\mathbf{E}|^2}{2} dV = \frac{1^3 \cdot \epsilon_0 \cdot 2^2}{2} = 2 \epsilon_0$$

30. For surface 1 below with D-field \mathbf{D} between 2 metal plates as shown, and for current "I" in the wire, then

$$\int_{\text{Surface 1}} \frac{\partial \mathbf{D}}{\partial t} \cdot d\mathbf{S} = I.$$

- a) True b) False

